Steady Sate Analysis of Self-Excited Induction Generator using Phasor-Diagram Based Iterative Model

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Abstract: – Induction generators in self excited mode are found to be suitable for remote and windy locations. Prior to installation there is a need to predict the behaviour of machine under all possible operating conditions. This is possible through steady state modeling of such generators. Steady state analysis of self excited induction generator (SEIG) needs an estimation of generated frequency and magnetizing reactance under all possible operating conditions. So far most of the researchers have used loop impedance, nodal admittance or iterative techniques to determine the steady state performance of such machines. In this paper a new model based upon phasor diagram of induction generator has been proposed to analyze the behaviour of self excited induction generator. Modeling results in a third order equation in generated frequency and a simple expression for magnetizing reactance. Complete mathematical analysis to derive the different expressions is presented here. Computed results have been compared with experimental results on test machines. Closeness between the two proves the validity of proposed modeling.

Keywords: - Non-conventional sources, Renewable energy, Steady state analysis, Self-excited induction generator (SEIG), Wind energy.

Nomenclature

- *a* per unit frequency
- *b* per unit speed
- *C* excitation capacitance per phase
- E_1 air gap voltage per phase at rated frequency
- *Err1* error in successive values of generated frequency
- *Err2* error in successive values of magnetizing reactance per phase
- *Err3* error in successive values of stator voltage per phase
- I_1 stator current per phase
- *I*₂ rotor current per phase, referred to stator

- I_C capacitor current per phase
- I_L load current per phase
- I_{rc} core loss current per phase
- I_m magnetizing current per phase
- *R* load resistance per phase
- R_1 stator resistance per phase
- R_c core loss resistance per phase
- *R*₂ rotor resistance per phase, referred to stator
- V_t stator voltage per phase
- *X* load reactance per phase
- X_1 stator reactance per phase

- *X*₂ rotor reactance per phase, referred to stator
- X_C capacitive reactance due to *C* at rated frequency
- X_m magnetizing reactance per phase at rated frequency

1. Introduction

A rapid increase in power demand and continuous depletion of fossil fuels has diverted the attention of scientists from conventional energy sources to nonconventional energy sources. Wind energy is emerging as an potential source among various non conventional energy sources. Most of the countries across the world are promoting such wind energy generating units.

Induction generators with cage rotor are found to be most suitable for wind energy conversion due to their advantages such as simple and rugged construction, low cost and no need of synchronization with existing grid. These machines can be operated in grid connected as well as in self-excited mode. Induction generator in self-excited mode is found to be capable to generate the power even in the absence of power grid. This makes it to be most useful machine for the remote windy locations.

Various methodologies adopted for the analysis of SEIG by researchers [1-9] are;

- Loop impedance technique
- Nodal admittance technique
- Iterative technique

Above methods require either lengthy derivations or solution of nonlinear equations. In this paper an attempt has been made to estimate the generated frequency and magnetizing reactance for self-excited induction generator using a new strategy based upon phasor diagram of the machine. Proposed modeling results in the third order equation in generated frequency and a simple expression for magnetizing reactance. Comparison of computed and experimental results on test machines confirms the validity of proposed modeling.

2. Steady-State Analysis

The steady-state operation of the selfexcited generator may be analyzed by using the equivalent circuit representation as shown in Fig.1. In this circuit all parameters are assumed to be constant except magnetizing reactance.



Fig.1 Equivalent circuit representation.

Analysis of Fig. 1 in the absence of any power source & with $V_t \& E_l$ as potential of node 1 &2 gives.

$$\frac{E_1}{jX_m} + \frac{aE_1}{R_c} + \frac{E_1 - \frac{V_t}{a}}{Z_1} + \frac{E_1}{\frac{R_2}{a-b} + jX_2} = 0$$

Where, $Z_1 = (R_1/a) + j X_1$

$$\frac{E_{1} - \frac{V_{t}}{a}}{Z_{1}} + \frac{\frac{V_{t}}{a}}{-j\frac{X_{c}}{a^{2}}} + \frac{\frac{V_{t}}{a}}{\frac{R}{a}} = 0$$
(2)

Equation (1) and (2) gives;

$$\frac{E_1}{jX_m} + \frac{aE_1}{R_c} + \frac{\frac{V_t}{a}}{-j\frac{X_c}{a^2}} + \frac{\frac{V_t}{a}}{\frac{R_1}{a}} + \frac{E_1}{\frac{R_2}{a-b} + jX_2} = 0 \quad (3)$$

Separation of real and imaginary parts of (3), results in the following;

$$\frac{\frac{E_1R_2}{a-b}}{\left(\frac{R_2}{a-b}\right)^2 + X_2^2} + \frac{aE_1}{R_c} + \frac{V_t}{R} = 0 \qquad (4)$$

and

$$-\frac{E_{1}X_{2}}{\left(\frac{R_{2}}{a-b}\right)^{2}+X_{2}^{2}}+\frac{aV_{t}}{X_{c}}-\frac{E_{1}}{X_{m}}=0$$
 (5)

$$X_{m} = \frac{E_{1}}{a\omega CV_{t} - \frac{E_{1}X_{2}}{\left(\frac{R_{2}}{a-b}\right)^{2} + X_{2}^{2}}}$$
(7)

Analysis of phasor diagram of induction generator as shown in Fig.2 gives;

$$V_{t} = a_{0}\sqrt{E_{1}^{2} - E_{1y}^{2}} - I_{1}R_{1}\cos\theta + a_{0}I_{1}X_{1}\sin\theta$$
-----(8)

Simplification of (4) gives a simple expression in *a* as;

$$A_{3}a^{3} + A_{2}a^{2} + A_{1}a + A_{0} = 0$$
 (6)

Where

$$A_{3} = RX_{2}^{2}E_{1}$$

$$A_{2} = -2bRX_{2}^{2}E_{1} + V_{t}R_{c}X_{2}^{2}$$

$$A_{1} = E_{1}RR_{2}^{2} + RE_{1}b^{2}X_{2}^{2} + R_{c}RE_{1}R_{2} - 2V_{t}R_{c}bX_{2}^{2}$$

$$A_{0} = V_{t}R_{c}R_{2}^{2} - R_{c}RE_{1}R_{2}b + V_{t}R_{c}b^{2}X_{2}^{2}$$

Solution of (6) gives the generated frequency for known values of operating speed and load resistance. Further exclusion of core loss branch leads to a quadratic equation in unknown frequency, in contrast to higher order polynomial equation in 'a' as obtained by other research persons.

Equation (5) gives the unknown magnetizing reactance as;



Fig.2. Phasor diagram of induction generator

Where

$$E_{Iy} = \frac{I_1 R_1 \sin \theta}{a} + I_1 X_1 \cos \theta,$$

$$I_1 = \sqrt{I_L^2 + I_C^2}$$

 E_{Iy} is resolved component of E_I along an axis perpendicular to terminal voltage.

3. Iteration Technique

Generated voltage and frequency for SEIG can be estimated using the following proposed iteration technique.

Step1. Computation of initial values of generated frequency and magnetizing reactance using following expression;

$$X_{m0} = \frac{1}{\frac{a_0}{X_c} - \frac{X_2(a_0 - b)^2}{R_2^2 + X_2^2(a_0 - b)^2}}$$

Initial values as;

$$E_{10} = V_{t0} = 1$$
 pu, $a_0 = 0.9999b$

Step2. Computation of air gap voltage, E_1 from magnetization characteristics (see Appendix-1 and Appendix-2).

Step3. Computation of modified values of stator voltage from (8).

Step4. Estimation of generated frequency '*a*' and magnetizing reactance ' X_m ' from (6) and (7) after using the modified values of E_1 and V_t .

Step5.Comparison of the new value of generated frequency 'a' with previous value i.e. a_0 as used in step 1.

If $ErrI = |a - a_0| \langle \varepsilon \rangle$, Where $\varepsilon = 0.000001$, *a* is treated as generated frequency and modified value of V_t may be treated as terminal voltage for SEIG.

If it is not so, process may be repeated by replacing ' a_0 ' with 'a' until difference in the successive values for generated frequency comes out to be ε .

It is also possible to compute the generated voltage and frequency by comparing the successive values of X_m and V_t obtained after each iteration. For such comparison iterative procedure terminates only if following expressions are satisfied.

$$Err2 = |X_m - X_{m0}| \langle \varepsilon$$

$$Err3 = \left| V_t - V_{t0} \right| \langle \mathcal{E}$$

4. Results and Discussions

Modeling proposed in the paper is found to be useful for estimation of generated frequency and magnetizing reactance of SEIG. Further, inclusion of core loss branch makes the analysis more realistic.

b=1.0133,Cpu=0.8448									
Sr. No.	Zpu	Experimental		Simulated Results using Phasor Diagram Based Analysis					
		Results		(PDA)					
		а	V _t	Errl		Err2		Err3	
				а	V_t	а	V_t	а	V_t
1.	2.3632	0.9820	0.9078	0.9820	0.9721	0.9820	0.9720	0.9820	0.9720
2.	2.5004	0.9830	0.9210	0.9835	0.9858	0.9835	0.9858	0.9835	0.9858
3.	2.6628	0.9856	0.9526	0.9851	1.0002	0.9851	1.0002	0.9851	1.0002
4.	2.8312	0.9864	0.9684	0.9866	1.0134	0.9866	1.0133	0.9866	1.0133
5.	3.0247	0.9870	0.9868	0.9881	1.0266	0.9881	1.0265	0.9881	1.0265
6.	3.5013	0.9892	1.0131	0.9912	1.0527	0.9912	1.0527	0.9912	1.0527
7.	4.4193	0.9922	1.0578	0.9953	1.0869	0.9953	1.0868	0.9953	1.0868
8.	6.1056	0.9960	1.0921	0.9997	1.1224	0.9997	1.1224	0.9997	1.1224
9.	6.6759	0.9968	1.1000	1.0007	1.1303	1.0007	1.1303	1.0004	1.1285
10.	7.1224	0.9984	1.1052	1.0013	1.1356	1.0013	1.1356	1.0013	1.1356

TABLE 1. COMPARISON OF EXPERIMENTAL AND SIMULATED RESULTS (MACHINE-1).

Table 1 shows the comparison of experimental and simulated results on Machine-1 [Appendix-1] using iterative process as explained in section-3.

It is observed that irrespective of error function (in terms of $a/X_m/V_t$) the final values for generated voltage and frequency turns out to be same up to 3rd digit after decimal.

However minimum numbers of iterations are required for final results in case error function is in terms of frequency. Table 2 shows the simulated results for Machine-2. The closeness between the experimental and simulated results as shown proves the validity of modeling adopted.

Simulated results may be obtained for any type of load connected across the machine

provided equations (6) and (7) are modified as given in appendix-2. Fig 3 and Fig 4 shows the variation of generated frequency and terminal voltage with load for different load power factors. Unity power factor load seems to be justified for better frequency and voltage regulation.



Fig. 3. Variation of generated frequency with load.

Fig. 4 Variation of stator voltage with load.

			Simulated Results						
Sr.	Exp	perimental R	using Phasor						
No.			Diagram Based						
				Analysis (PDA)					
	Ь	а	V _t	a	V_t				
Zpu=3.4542, Cpu=36									
1.	0.9533	0.9438	0.5826	0.942680	0.6644				
2.	0.9780	0.9660	0.6869	0.967044	0.7675				
3.	0.9986	0.9870	0.7652	0.987454	0.8400				
4.	1.0106	0.9984	0.8217	0.999305	0.8818				
5.	1.0286	1.0148	0.8826	1.017080	0.9440				
6.	1.0466 1.0338		0.9434	1.034854	0.9975				
7.	1.0640 1.0508		0.9913	1.051969	1.0371				
<i>Zpu</i> =3.4542, <i>Cpu</i> =51									
8.	0.8533	0.8434	0.7217	0.8437	0.7605				
9.	0.8806	0.8700	0.8130	0.8707	0.8434				
10.	0.9020	0.8934	0.8739	0.8917	0.8920				
11.	0.9266	0.9182	0.9347	0.9161	0.9479				
12.	0.9600	0.9480	1.0086	0.9490	1.0231				
<i>Zpu</i> =4.7495, <i>Cpu</i> =36									
13.	0.9613	0.9528	0.6695	0.9534	0.7417				
14.	0.9780	0.9656	0.7434	0.9699	0.8010				
15.	0.9973 0.9848		0.8173	0.9891	0.8692				
16.	1.0266 1.0156		0.9130	1.0181	0.9714				
17.	1.0420	1.0226	0.9739	1.0333	1.0095				
<i>Zpu</i> =4.7495, <i>Cpu</i> =51									
18.	0.8566	0.8486	0.7565	0.8495	0.7981				
19.	0.8766	0.8682	0.8130	0.8694	0.8521				
20.	0.9000	0.8900	0.8782	0.8925	0.9057				
21.	0.9240	0.9140	0.9391	0.9163	0.9605				
22.	0.9373	0.9282	0.9695	0.9295	0.9909				
23.	0.9533	0.9524	1.0304	0.9453	1.0272				

TABLE 2:- COMPARISON OF EXPERIMENTAL AND SIMULATED RESULTS (MACHINE-2)

5. Conclusions

In this paper attempt has been made to estimate the steady-state performance of self-excited induction generator using a new iterative technique approach based upon phasor diagram of the machine. The proposed technique has not been used by any other research person so far. Iterative technique is found to be very simple and

effective. Simulated results are verified using experimental results on two test machines with different ratings. Close agreement between simulated and experimental results proves the validity of proposed modeling.

In future this research work may be extended to estimate the excitation capacitance and operating speed to obtain a constant voltage constant frequency operation. This may be helpful to promote the applications of self excited induction generators in remote and windy areas. In turn it may be helpful to preserve the conventional fuels which are likely to be finished with time.

Appendix 1

The details of the induction Machine-1 used to obtain the experimental results are;

• Specifications

3-phase, 4-pole, 50 Hz, star connected, squirrel cage induction machine 750W/1HP, 380 V, 1.9 A

• Parameters The equivalent circuit parameters for the machine in pu are

 $R_1 = 0.0823, R_2 = 0.0696, X_1 = X_2 = 0.0766$

• Base values Base voltage =219.3 V Base current =1.9 A Base impedance=115.4 Ω Base frequency=50 Hz Base speed=1500rpm

• Air gap voltage

The variation of magnetizing reactance with air gap voltage at rated frequency for the induction machine is as given below.

X_m (169.2	$E_1 = 512.69 - 2.13X_m$
$179.42\rangle X_m \geq \! 169.20$	$E_1 = 891.66 - 4.37 X_m$
$184.46\rangle X_m \geq \! 179.42$	$E_1 = 785.79 - 3.78X_m$
$X_m \ge 184.46$	$E_1 = 0$

Appendix 2

The details of the induction Machine-2 used to obtain the experimental results are;

• Specifications 3-phase, 4-pole, 50 Hz, delta connected, squirrel cage induction machine 2.2kW/3HP, 230 V, 8.6 A • Parameters

The equivalent circuit parameters for the machine in pu are

$$R_1 = 0.0723, R_2 = 0.0379, X_1 = X_2 = 0.1047$$

• Base values Base voltage =230 V Base current =4.96 A Base impedance=46.32 Ω Base frequency=50 Hz Base speed=1500rpm

• Air gap voltage

The variation of magnetizing reactance with air gap voltage at rated frequency for the induction machine is as given below.

$$\begin{array}{ll} X_m \langle 82.292 & E_1 = 344.411 - 1.61 X_m \\ \\ 95.569 \rangle X_m \geq 82.292 & E_1 = 465.12 - 3.077 X_m \\ 108.00 \rangle X_m \geq 95.569 & E_1 = 579.897 - 4.278 X_m \\ \\ X_m \geq 108.00 & E_1 = 0 \end{array}$$

Appendix-III

Case I: For R load (excluding R_c)

Expression in *a* can be written as;

$$A_2a^2 + A_1a + A_0 = 0$$

Where

$$A_2 = V_t X_2^2$$

 $A_1 = E_1 R R_2 - 2bV_t X_2^2$
 $A_0 = b^2 V_t X_2^2 - bE_1 R R_2 + V_t R_2^2$

The expression of magnetizing reactance can be written as;

$$X_m = \frac{E_1}{a\omega CV_t - \frac{E_1X_2}{\left(\frac{R_2}{a-b}\right)^2 + X_2^2}}$$

Here stator current, I_1 expression is same as written in the IV section.

Case II: For *RL* load (including *R_c*)

Expression in *a* can be written as;

$$A_5a^5 + A_4a^4 + A_3a^3 + A_2a^2 + A_1a + A_0 = 0$$

Where;

$$A_{5} = E_{1}X^{2}X_{2}^{2}$$

$$A_{4} = -2E_{1}bX^{2}X_{2}^{2}$$

$$A_{3} = E_{1}R_{2}R_{c}X^{2} + E_{1}R^{2}X_{2}^{2} + X^{2}R_{2}^{2}E_{1} + E_{1}X^{2}X_{2}^{2}$$

$$A_{2} = -E_{1}R_{2}R_{c}bX^{2} - 2bX_{2}^{2} + V_{t}RR_{c}X_{2}^{2}$$

$$A_{1} = E_{1}R_{2}R_{c}R^{2} + E_{1}R^{2}R_{2}^{2} + E_{1}b^{2}X_{2}^{2}R^{2} - 2V_{t}RR_{c}bX_{2}^{2}$$

$$A_{0} = -E_{1}R_{2}R_{c}R^{2}b + V_{t}RR_{c}R_{2}^{2} + V_{t}RR_{c}b^{2}X_{2}^{2}$$

$$X = R\sqrt{\left(\frac{1}{pf}\right)^{2} - 1}$$

The expression of magnetizing reactance can be written as;

$$X_{m} = \frac{E_{1}}{a\omega CV_{t} - \frac{E_{1}X_{2}}{\left(\frac{R_{2}}{a-b}\right)^{2} + X_{2}^{2}} - \frac{\frac{V_{t}X}{a}}{\left(\frac{R}{a}\right)^{2} + X^{2}}}$$

Case III: For *RL* load (excluding *R_c*)

Expression in *a* can be written as;

$$A_3a^3 + A_2a^2 + A_1a + A_0 = 0$$

Where;

$$A_{3} = E_{1}R_{2}X^{2}$$

$$A_{2} = V_{t}RX_{2}^{2} - E_{1}R_{2}bX^{2}$$

$$A_{1} = -2bV_{t}RX_{2}^{2} + E_{1}R_{2}R^{2}$$

$$A_{0} = V_{t}RR_{2}^{2} + b^{2}V_{t}RX_{2}^{2} - E_{1}R_{2}R^{2}b$$

$$X = R\sqrt{\left(\frac{1}{pf}\right)^{2} - 1}$$

The expression of magnetizing reactance can be written as;

$$X_{m} = \frac{E_{1}}{a\omega CV_{t} - \frac{E_{1}X_{2}}{\left(\frac{R_{2}}{a-b}\right)^{2} + X_{2}^{2}} - \frac{\frac{V_{t}X}{a}}{\left(\frac{R}{a}\right)^{2} + X^{2}}}$$

Here stator current, I_1 expression for case III and case IV, is modified and written as;

$$I_1 = \sqrt{\left(I_C - I_L \sin \varphi\right)^2 + \left(I_L \cos \varphi\right)^2}$$

Case IV: For *RC* load (including *R_c*)

Expression in *a* can be written as;

$$A_5a^5 + A_4a^4 + A_3a^3 + A_2a^2 + A_1a + A_0 = 0$$

Where;

$$A_{5} = E_{1}R^{2}X^{2}$$

$$A_{4} = V_{t}RR_{c}X_{2}^{2} - 2bE_{1}R^{2}X_{2}^{2}$$

$$A_{3} = -2bV_{t}RR_{c}X_{2}^{2} + E_{1}R^{2}R_{2}^{2} + b^{2}E_{1}X_{se}^{2}X_{2}^{2} + E_{1}R_{2}R_{c}R^{2}$$

$$A_{2} = V_{t}RR_{c}R_{2}^{2} + b^{2}V_{t}RR_{c}X_{2}^{2} - 2bE_{1}X_{se}^{2}X_{2}^{2} - E_{1}R_{2}R_{c}R^{2}b$$

$$A_{1} = E_{1}X_{se}^{2}R_{2}^{2} + b^{2}E_{1}X_{se}^{2}X_{2}^{2} + E_{1}R_{2}R_{c}X_{se}^{2}$$

$$A_{0} = -E_{1}R_{2}R_{c}X_{se}^{2}b$$

$$X_{se} = aR\sqrt{\left(\frac{1}{pf}\right)^{2} - 1}$$

The expression of magnetizing reactance can be written as;

$$X_{m} = \frac{E_{1}}{\frac{aV_{t}}{X_{c}} - \frac{E_{1}X_{2}}{\left(\frac{R_{2}}{a-b}\right)^{2} + X_{2}^{2}} - \frac{aV_{t}X_{se}}{\left(aR\right)^{2} + X_{se}^{2}}}$$

Case V: For RC load (excluding R_c)

Expression in *a* can be written as;

$$A_4a^4 + A_3a^3 + A_2a^2 + A_1a + A_0 = 0$$

Where;

$$A_{4} = V_{t}RX_{2}^{2}$$

$$A_{3} = -2V_{t}RX_{2}^{2} + E_{1}R_{2}R^{2}$$

$$A_{2} = V_{t}RR_{2}^{2} + b^{2}V_{t}RX_{2}^{2} - E_{1}R_{2}bR^{2}$$

$$A_{1} = E_{1}R_{2}X_{se}^{2}$$

$$A_{0} = -bE_{1}R_{2}X_{se}^{2}$$

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