Optimal Control of Flying Objects' Move After Estimated State Vector Using a Reduced Order Observer

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Abstract: - In this paper one presents an algorithm for project of a reduced order observer (ALGLOOR) with applications to flying objects' move. The command law is an optimal one in rapport with state vector of the observer. Numeric calculus examples are also presented: for longitudinal and lateral move and vertical move of a rocket in vertical plane in rapport with equal signal line; using Matlab/Simulink model of closed loop system one has made numeric calculus programs and with them one obtained different time characteristics, which expresses state variables' dynamics of the flying objects dynamic models.

Key-Words: - algorithm, optimal, control law, rocket, state vector, reduced order observer, command

1 Introduction

Observers are used for estimation of flying object state vector using its input and output signals. Measuring some of the state vector's components (output vector's components) one reduces sensors' number especially the number of sensors for reachless or difficult measurable state variables (e.g. elastic deformations) [1], [2], [3]. First observer structures have been developed by Luenbergher in 1966, 1971, 1979. The observers are used both for systems with known inputs and unknown inputs systems; their project belongs to different researchers (Bhatta-Charyya, 1978; Chen and Patton, 1999; Darouach, 1994; Hostetter and Meditch, 1973; Hou and Müller, 1992; Hui and Zak, 1993 and 2005; Kudva, 1980; Kurek, 1983; Wang, 1975; Yang and Wilde, 1988; Krzminski and Kakzorek, 2005 and so on). Recent observers are

obtained for systems with some unknown inputs or for systems with subsystems whose dynamic is unknown. For such systems one may use reduced observers. Unknown inputs order may he disturbances, errors and so on.

The estimator (filter) Kalman – Bucy [4], [5], [6] is the best observer for stochastic systems' state estimation. In case of nonlinear systems with partial known inputs or with unknown subsystems adaptive observers have been made; these observers use neural networks [7], [8], [9], [10], [11].

Recent researches deal with project of observers for the intern delay systems [12], [13], [14], based on Liapunov theory; observers' project means matrices inequalities solve. Linear state observers are used for x state estimation for linear system using measured variables vector y, the input vector or some components of this vector.

2 Dynamic models of flying objects and reduced order observer

Flying objects' dynamic may be described by equations [15]

$$\dot{x} = Ax + Bu + Eu_p, \qquad (1)$$

$$y = Cx, (2)$$

where x is the state vector $(n \times 1)$ of aircraft model, y – output vector $(p \times 1), u$ – the vector containing knowable inputs $(m_1 \times 1), u_p$ - vector $(m_2 \times 1)$ of unknown inputs (disturbances or nonlinear functions of non modeled systems) [16], A- matrix $(n \times n), B - \text{matrix} (n \times m_1), E - \text{matrix} (n \times m_2),$ C – matrix $(p \times n)$; matrices A, B, C are known.

Let's consider the reduced order observer described by equations [17], [18]

$$M\dot{z} = Fz + Gu + Hy, \tag{3}$$

$$\hat{x} = Pz + Qy, \tag{4}$$

with $z(r \times 1), \hat{x}(n \times 1), M(r \times r), F(r \times r), \det M = 0$ and rang $M \le n - q$, $G(r \times m)$, $H(r \times p)$, $P(n \times r)$, a is the ch Q(i

$$n \times p$$
); $m = m_1 + m_2$; e is the observer's error

$$e = z - Nx, (5)$$

By derivation of equation (5), replacing \dot{x} with form (1), y with form (2), \dot{z} with form (3) and assenting that all coefficients of x, u and u_p be null,

$$G = MNB, (6)$$

$$HC = MNA - FN, (7)$$

$$MNE = 0, (8)$$

one obtains the equation of error e

$$M\dot{e} = Fe. \tag{9}$$

(10)

Error $(\hat{x} - x)$ may be expressed in rapport with e. Taking into account equations (4) and (5), one results

 $\hat{x} - x = Pe$

if

$$PN + QC = I \Leftrightarrow \begin{bmatrix} P & Q \end{bmatrix} \begin{bmatrix} N & C \end{bmatrix}^T = I.$$
(11)
unity matrix

I - unity matrix.

3 Project algorithm of the reduced order observer (ALGLOOR)

First of all one chooses matrix N of form

$$N = \begin{bmatrix} N_1 & N_2 \\ N_3 & N_4 \end{bmatrix} = \begin{bmatrix} 0_{(n-q)\times(n-p)} & I_{(n-q)\times p} \\ 0_{(n-p)\times(n-p)} & 0_{(n-p)\times p} \end{bmatrix}, \quad (12)$$

where $I_{(n-p) \times (n-p)} = I_{n-p}$ is unity matrix and $I_{(n-q) \times p}$ is a matrix whose elements are equal with 1 at intersection of line *i* and column j = i and all the other elements are null; p = rang C, q = rang E= n - p.

One calculates matrix M so that condition (8) be fulfilled. For this, matrix M is partitioned

$$M = \begin{bmatrix} M_{1(n-p) \times (n-q)} & M_{2(n-p) \times (n-p)} \\ M_{3(n-q) \times (n-q)} & M_{4(n-q) \times (n-p)} \end{bmatrix}.$$
 (13)

With E of form

$$E^{T} = \begin{bmatrix} E_{1} & E_{2} \end{bmatrix}, E_{1} \begin{bmatrix} (n-p) \times q \end{bmatrix}, \\ E_{2} \begin{bmatrix} (n-q) \times q \end{bmatrix},$$
(14)

equation (8) is equivalent with system

$$M_{2}E_{1} + M_{1}E_{2} = 0,$$

$$M_{4}E_{1} + M_{3}E_{2} = 0.$$
(15)

Choosing

$$M_{2} = 0_{(n-p) \times [n-p]}, M_{4} = 0_{(n-q) \times (n-p)},$$

one results

$$M = \begin{bmatrix} M_1 & 0 \\ M_3 & 0 \end{bmatrix},$$
 (16)
$$M_1 E_2 = 0, M_3 E_2 = 0.$$

With M and N calculated, one determines matrix G using equation (6). Also, one calculates matrices F and H with (7). For this, matrix F is partitioned as follows

$$F = \begin{bmatrix} F_{1(n-p)\times(n-q)} & 0_{(n-p)\times(n-p)} \\ F_{3(n-q)\times(n-q)} & 0_{(n-q)\times(n-p)} \end{bmatrix}.$$
 (17)

With H and A of forms

$$H^{T} = \begin{bmatrix} H_{1} & H_{2} \end{bmatrix}, H_{1} \begin{bmatrix} (n-p) \times p \end{bmatrix}, H_{2} \begin{bmatrix} (n-q) \times p \end{bmatrix}, \quad (18)$$

$$A = \begin{bmatrix} A_{1(n-p)\times(n-p)} & A_{2(n-p)\times p} \\ A_{3p\times(n-p)} & A_{4p\times p} \end{bmatrix},$$
 (19)

condition (7) is equivalent with system

$$M_{1}A_{3} = H_{1}C_{1},$$

$$M_{1}A_{4} - F_{1} = H_{1}C_{2},$$

$$M_{3}A_{3} = H_{2}C_{1},$$

$$M_{2}A_{4} - F_{2} = H_{2}C_{2};$$
(20)

one results

$$H_{1} = M_{1}A_{3}C_{1}^{+},$$

$$H_{2} = M_{3}A_{3}C_{1}^{+},$$
(21)

$$F_1 = M_1 A_4 - M_1 A_3 C_1^+ C_2, \qquad (22)$$

$$F_3 = M_3 A_4 - M_3 A_3 C_1^+ C_2 \,. \tag{(--)}$$

One yield

$$H = \begin{bmatrix} M_{1}A_{3}C_{1}^{+} \\ M_{3}A_{3}C_{1}^{+} \end{bmatrix},$$

$$F = \begin{bmatrix} M_{1}A_{4} - M_{1}A_{3}C_{1}^{+}C_{2} & 0_{(n-p)\times(n-p)} \\ M_{3}A_{4} - M_{3}A_{3}C_{1}^{+}C_{2} & 0_{(n-q)\times(n-p)} \end{bmatrix}.$$
(23)

Matrices P and Q are obtained from equation (11). These matrices are partitioned as follows

$$P = \begin{bmatrix} P_{1(n-p) \times (n-q)} & P_{2(n-p) \times (n-p)} \\ P_{3(n-q) \times (n-q)} & P_{4(n-q) \times (n-p)} \end{bmatrix},$$

$$Q = \begin{bmatrix} Q_{1(n-p) \times p} \\ Q_{2(n-q) \times p} \end{bmatrix}.$$
(24)

Condition (11) is equivalent with system

$$P_{2} + Q_{1}C_{1} = I_{(n-p) \times (n-p)},$$

$$P_{1} + Q_{1}C_{2} = 0_{(n-p) \times (n-q)},$$
(25)

$$P_4 + Q_2 C_1 = 0_{(n-q) \times (n-p)},$$

 $P_4 + Q_2 C_1 = I_1$

$$f_3 + Q_2 C_2 = I_{(n-q)\times(n-q)}$$
.

For example, one chooses

$$P_{1} = 0_{(n-p)\times(n-q)},$$

$$P_{3} = 0_{(n-q)\times(n-q)}$$
(26)

and equations (26) become

$$P_2 + Q_1 C_1 = 0, Q_1 C_2 = 0; (27)$$

$$P_4 + Q_2 C_1 = 0, Q_2 C_2 = I; \qquad (28)$$

$$Q_2 = C_2^+, P_4 = -C_2^+ C_1, \tag{29}$$

and Q_1 and P_2 represent solution of system (28).

The above results lead to following algorithm (ALGLOOR).

Step 1: one calculates $p = \operatorname{rang} C, q = \operatorname{rang} E$ and verifies condition p + q = n; if this condition isn't fulfilled then the algorithm isn't useful in order to project the reduced order observer.

Step 2: one calculates matrix N with (12) and matrix M with (16);

Step 3: one calculates matrix G with (6);

Step 4: matrices *A*, *C* are partitioned; one calculates C_1^+ and, using this, matrices *H* and *F* are calculated with equation (23);

Step 5: matrices P,Q are partitioned using (24) and after that one calculates their components with (27), (28) and (30).



Fig.1 Modeling diagram for optimal command system

In fig.1 modeling diagram for optimal command system described by equations (1) and (2) and which has as component a reduced order observer (equations (3) and (4)) is presented.

One remarks that the algorithm ALGLOOR doesn't use non singular transformations like other algorithms [19], [20], [21].

4 Numeric examples

Example 4.1 (aircraft longitudinal move)

A reduced order observer for state \hat{x} estimation described by equation (31) from [1] for longitudinal move with

$$C = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix}, E^{T} = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix},$$

$$n = 4, p = 2, q = 2$$

may be projected using ALGLOOR algorithm

$$\begin{bmatrix} \Delta \dot{V}_{x} \\ \Delta \dot{\alpha} \\ \Delta \dot{\theta} \\ \Delta \dot{\omega}_{y} \end{bmatrix} = \begin{bmatrix} -0.007 & 0.012 & -9.81 & 0 \\ -0.128 & -0.54 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0.065 & 0.96 & 0 & -0.99 \end{bmatrix} \begin{bmatrix} \Delta V_{x} \\ \Delta \alpha \\ \Delta \theta \\ \Delta \omega_{y} \end{bmatrix} + \begin{bmatrix} 0 \\ -0.04 \\ 0 \\ -12.5 \end{bmatrix} \delta_{p}.$$
 (31)

Matlab program utilized is the first presented in Section 4; one obtained matrices

$$N = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix}, M = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix}, G = \begin{bmatrix} 0 \\ -12.5 \\ 0 \\ -12.5 \\ 0 \\ -12.5 \end{bmatrix},$$

$$H = \begin{bmatrix} 0 & 0 \\ 0.065 & 0.065 \\ 0 & 0 \\ 0.065 & 0.065 \end{bmatrix}, F = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & -0.99 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & -0.99 & 0 & 0 \end{bmatrix},$$

$$P = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix}, Q = \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 0 \\ 0 & 0 \end{bmatrix}.$$
Fig 2 Matlab/Simulink model of the

rig.2 Matiab/Simulink model of the closed loop system

In fig.2 Matlab/Simulink model of the closed loop system described by equations $(1) \div (4)$ and $u = -K\hat{x}$ is presented. For u = 0 (open loop) one obtains characteristics $x_i(t)$ (blue) and $\hat{x}_i(t)$ (red)-fig.3. These characteristics are also obtained for closed loop system – fig.4.



Fig.3 Characteristics $x_i(t)$ and $\hat{x}_i(t)$ (open loop)



Fig.4 Characteristics $x_i(t)$ and $\hat{x}_i(t)$ (closed loop)

ALGLOOR may be used with good results using longitudinal move's model from [22]

Example 4.2 (aircraft lateral move)

Similar, for lateral move described by equation [1] -0.0558 -0.9968 0.0802 0.0415 Δβ 0.0073 (33)Δβ 0 -0 598 -0.115 -0.0318Δώ, 0 Δω -0.475 0.123 δ. 0.305 0.388 -0.4650 $\Delta \omega_x$ 0.153 1.063 δ, Δώ 0 0.0805 1 0 0 0 ΔφJ Δφ are known

$$C = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix}, E^{T} = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix},$$

$$n = 4, \ n = a = 2;$$

one obtains

$$N = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix}, M = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix}, G = \begin{bmatrix} 0.15 & 1.06 \\ 0 & 0 \\ 0.15 & 1.06 \\ 0 & 0 \end{bmatrix}, (34)$$
$$H = \begin{bmatrix} 0.30 & 0.38 \\ 0 & 0.08 \\ 0.30 & 0.38 \\ 0 & 0.08 \end{bmatrix}, F = \begin{bmatrix} -0.46 & 0 & 0 & 0 \\ 1 & 0 & 0 & 0 \\ -0.46 & 0 & 0 & 0 \\ 1 & 0 & 0 & 0 \end{bmatrix}, Q = \begin{bmatrix} 0.46 & 0 & 0 & 0 \\ -0.46 & 0 & 0 & 0 \\ 1 & 0 & 0 & 0 \end{bmatrix}, P = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 1 & 0 & 0 & 0 \end{bmatrix}, Q = \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 0 \\ 0 & 0 \end{bmatrix}.$$

Characteristics $x_i(t)$ (blue) and $\hat{x}_i(t)$ (red) are presented in fig.5 (without controller) and fig.6 (with controller $u = -K\hat{x}$).



Fig.5 Characteristics $x_i(t)$ and $\hat{x}_i(t)$ (without controller)



(with controller)

Example 4.3 (rocket move)

One considers the case of a rocket move in vertical plane; it is directed by three points method (the coliniarity CP–R–T; CP – control point, R – rocket, T - target). Lateral deviation z in rapport with equal signal line is described by differential equation [23], [24]

$$\dot{z} = V \vartheta + V f_{T}, \qquad (35)$$

where V is the flight velocity, ϑ – slope of the trajectory ($\vartheta = \theta - \alpha, \theta$ – pitch angle), f_T – disturbance. ϑ is defined by [24]

$$T_V \dot{\vartheta} = \alpha - T_V \frac{g}{V} \cos \vartheta.$$
 (36)

Expressing the above equation in variables α and θ , it becomes

$$T_{v}\dot{\theta} = T_{v}\dot{\alpha} + \alpha - T_{v}\frac{g}{V}\cos\vartheta$$
(37)

or has the liniarised form

$$T_{V}\dot{\theta} = T_{V}\Delta\dot{\alpha} + \Delta\alpha - \tilde{\alpha},$$

$$\tilde{\alpha} = -\frac{g}{V}T_{V}\cos\vartheta_{0}, \dot{\theta} = \Delta\dot{\theta};$$
(38)

Flying object's move in vertical plane is defined by equation (36) and the following one

$$\dot{V} + \frac{c_x}{c_y^{\alpha} T_V} V = \left(1 + \frac{c_x}{c_y^{\alpha}}\right) \frac{F_T}{m} - g\sin\vartheta \qquad (39)$$

Rocket's move around mass centre in vertical plane is described by equation [24]

$$J_z \theta + m_{\dot{\theta}} \theta + m_{\alpha} \alpha = m_{\delta} \delta, \qquad (40)$$

where J_z is the inertia moment in rapport with lateral (horizontal) axis, m_{θ} – dynamic damp moment coefficient, m_{α} – static stabilisation moment coefficient, m_{δ} – command coefficient moment. With $\theta = \vartheta + \alpha$, (36) and (39) the above equation leads to linear form

$$\Delta \ddot{\alpha} + 2\xi \omega_0 \Delta \dot{\alpha} + \omega_0^2 \Delta \alpha = k_\delta \delta + m_{\dot{\theta}} \frac{g}{J_z V} \cos \theta_0, \quad (41)$$

 $\Delta \delta = \delta,$

where ξ is the damp coefficient and ω_0 – proper frequency of the rotation move in vertical plane;

$$2\xi\omega_{0} = \frac{m_{\dot{\theta}}}{J_{z}} + \frac{1}{T_{v}}, k_{\delta} = \frac{m_{\delta}}{J_{z}},$$

$$\omega_{0}^{2} = \frac{m_{\alpha}}{J_{z}} + \frac{m_{\dot{\theta}}}{J_{z}T_{v}} + g \frac{\sin\theta_{0}}{VT_{v}} - \frac{\dot{T}_{v}}{T_{v}^{2}} \cong \frac{m_{\alpha}}{J_{z}} + \frac{m_{\dot{\theta}}}{J_{z}T_{v}}.$$
(42)

Equation (40) is equivalent with the following one

$$\ddot{\theta} = -\frac{m_{\dot{\theta}}}{J_z}\dot{\theta} - \frac{m_{\alpha}}{J_z}\alpha + k_{\delta}\delta, \qquad (43)$$

which, taking into account

$$\frac{m_{\dot{\theta}}}{J_{z}} = 2\xi \omega_{0} - \frac{1}{T_{v}} = a_{1},$$

$$\frac{m_{\alpha}}{J_{z}} \cong \omega_{0}^{2} - \frac{1}{T_{v}} \frac{m_{\dot{\theta}}}{J_{z}} = \omega_{0}^{2} - \frac{a_{1}}{T_{v}} = a_{0},$$
(44)

leads to

$$\ddot{\theta} = a_1 \dot{\theta} - a_0 \Delta \alpha + k_\delta \delta. \tag{45}$$

By derivation of equation (35), one obtains [25]

$$\ddot{z} = \frac{V}{T_v} \Delta \alpha + W_T, W_T = \dot{V}f_T + \dot{V}f_T - g\cos\vartheta_0; \quad (46)$$

 W_{T} is an equivalent disturbance having the significance of normal acceleration to equal signal line [26]. For calculus of above equations' coefficients one uses calculus equations from [27]

$$T_{V} = \frac{1}{d_{1}}, \omega_{0} = \sqrt{d_{1}d_{4} - d_{3}},$$

$$\xi = \frac{d_{1} + d_{4}}{2\sqrt{d_{1}d_{4} - d_{3}}}, k_{\delta} = d_{2},$$
(47)

where d_1, d_2, d_3, d_4 are read using diagrams or graphic characteristics for different rocket types at different flight seconds. For instance, for an Oerlikon rocket, in the 10th flight second $d_1 = 1.5 \text{ s}^{-1}, d_2 = 40 \text{ s}^{-2}, d_3 = -10 \text{ s}^{-2}, d_4 = 1.2 \text{ s}^{-1},$ V = 400 m/s, $T_V = 0.66 \text{ s}$, $\omega_0 = 4.669 \text{ s}^{-1}$, $\xi = 0.062$, (48) $k_{\delta} = d_2 = 40 \,\mathrm{s}^{-2}, a_1 = 0.92 \,\mathrm{s}^{-1}, a_0 = 23.18 \,\mathrm{s}^{-2}.$ For the 40th flight second $d_1 = 2.5 \,\mathrm{s}^{-1}, d_2 = 100 \,\mathrm{s}^{-2}, d_3 = -80 \,\mathrm{s}^{-2}, d_4 = 1 \,\mathrm{s}^{-1},$ $V = 400 \text{ m/s}, T_v = 0.4 \text{ s}, \omega_0 = 9.083 \text{ s}^{-1}, \xi = 0.0212, (49)$ $k_{\delta} = d_2 = 100 \,\mathrm{s}^{-2}, a_1 = 2.115 \,\mathrm{s}^{-1}, a_0 = 87.7881 \,\mathrm{s}^{-2}.$ For the 50th flight second $d_1 = 1 s^{-1}, d_2 = 25 s^{-2}, d_3 = -60 s^{-2}, d_4 = 0.4 s^{-1},$ $V = 400 \text{ m/s}, T_v = 1 \text{ s}, \omega_0 = 7.77 \text{ s}^{-1}, \xi = 0.011,$ (50) $k_{\delta} = d_2 = 25 \,\mathrm{s}^{-2}, a_1 = 0.8197 \,\mathrm{s}^{-1}, a_0 = 61.2082 \,\mathrm{s}^{-2}.$ Choosing the state vector $x^T = \begin{vmatrix} z & \dot{z} & \Delta \alpha & \dot{\theta} \end{vmatrix}$, system formed by equations (38), (45) and (46)

becomes $\dot{x}_1 = x_2,$ $\dot{x}_2 = \frac{V}{T_V} x_3 + W_T,$ (51)

$$\dot{x}_{3} = -\frac{1}{T_{v}}x_{3} + x_{4} + \frac{1}{T_{v}}\tilde{\alpha},$$

$$\dot{x}_{4} = -a_{0}x_{3} - a_{1}x_{4} + k_{\delta}\delta.$$
(51)

If the input is $u = \delta$ and the disturbances vector is $u_p^T = \begin{bmatrix} \widetilde{\alpha} & W_T \end{bmatrix}$, the above equations system has the form

$$\dot{x} = Ax + Bu + Eu_p, \tag{52}$$

where

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & \frac{V}{T_{v}} & 0 \\ 0 & 0 & -\frac{1}{T_{v}} & 1 \\ 0 & 0 & -a_{0} & -a_{1} \end{bmatrix}, B = \begin{bmatrix} 0 \\ 0 \\ 0 \\ k_{\delta} \end{bmatrix}, E = \begin{bmatrix} 0 & 0 \\ 0 & 1 \\ \frac{1}{T_{v}} & 0 \\ 0 & 0 \end{bmatrix}.$$
 (53)

Using ALGLOOR for the 10^{th} flight second, one obtains

$$C = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}, N = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 1 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}, G = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 0 \\ 0 \end{bmatrix}, H = \begin{bmatrix} 0 & 0 \\ 600 & 0 \\ 600 & 0 \\ 600 & 0 \\ 600 & 0 \end{bmatrix},$$

$$F = \begin{bmatrix} 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \end{bmatrix}, P = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}, Q = \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 1 & 0 \\ 0 & 1 \end{bmatrix}.$$
(54)

Eigen values of the open loop system are

$$s_1 = 0, s_2 = 0,$$

$$s_3 = 4.82, s_4 = -4.52.$$
(55)

In closed loop system one obtains the characteristics from fig.7.



Fig.7 State variables $x_i(t)$ and $\hat{x}_i(t)$ of the optimal control system for rocket move in vertical plane using 10th flight second parameters Similarly, for the 40th and 50th flight second, it results

$$C = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}, N = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 1 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}, G = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 0 \\ 0 \end{bmatrix}, H = \begin{bmatrix} 0 & 0 \\ 1000 & 0 \\ 0 & 0 \\ 1000 & 0 \end{bmatrix}, (56)$$

$$F = \begin{bmatrix} 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \end{bmatrix}, P = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}, Q = \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 1 & 0 \\ 0 & 1 \end{bmatrix}, et{tabular}$$

respectively,

$$C = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}, N = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 1 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}, G = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 0 \end{bmatrix}, H = \begin{bmatrix} 0 & 0 \\ 400 & 0 \\ 0 & 0 \\ 400 & 0 \end{bmatrix},$$

$$F = \begin{bmatrix} 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \end{bmatrix}, P = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}, Q = \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 1 & 0 \\ 0 & 1 \end{bmatrix}$$
Constant of the equation of th

and characteristics from fig.8 anf fig.9.



Fig.8 State variables $x_i(t)$ and $\hat{x}_i(t)$ of the optimal control system for rocket move in vertical plane using 40th flight second parameters



Fig.9 State variables $x_i(t)$ and $\hat{x}_i(t)$ of the optimal control system for rocket move in vertical plane using 50th flight second parameters

5 Matlab programs

One presents the program for the longitudinal move of the flying object and for rocket move in vertical plane at 10th flight second. The programs for lateral move or for rocket move in vertical plane at 40th and 50th flight second are similar. In programs presented in Subsections 5.1 and 5.2 one runs other subprograms: ALGLX & ALGLX10. These programs are shown in Subsections 5.3 and 5.4.

5.1 ALGLOOR program (longitudinal move)

clear all; run ALGLX; close all; clear C; clear contor; clear e; clear i; clear r; clear m; clear P; clear Q; KK=K; C=[1 0 0 0;0 1 0 0];

E = [1 0; 0 1; 0 0; 0 0];n=size(A,1); m=size(B,2); q=size(E,2); p=size(C,1); x0=[100;1;0;10]; xc0=[10;0;-10;20]; up=[-0.0375;-0.0963]; deltap=[5]; A1=A(1:(n-p),1:(n-p)); A2=A(1:(n-p),(n-p+1):size(A,2));A3=A((n-p+1):size(A,1),1:(n-p)); A4=A((n-p+1):size(A,1),(n-p+1):size(A,2));C1=C(:,1:(n-p));C2=C(:,(n-p+1):n); % Step 1 N1=zeros((n-q),(n-p)); N2=eye((n-q),p); N3=eye((n-p),(n-p)); N4=zeros((n-p),p); N=[N1 N2;N3 N4]; % Step 2 M1=eye((n-p),(n-q));M2=zeros((n-p),(n-p));M3=eye((n-q),(n-q)); M4=zeros((n-q),(n-p));M=[M1 M2;M3 M4]; % Step 3 I1=eye((n-p),(n-p)); I2=zeros((n-p),(n-q));I3=zeros((n-q),(n-p)); I4=eye((n-q),(n-q));I=[I1 I2;I3 I4]; P1=zeros((n-p),(n-q));P2=zeros((n-p),(n-p));P3=eye((n-q),(n-q));P4=-(pinv(c2))*c1;P=[P1 P2;P3 P4]; Q1=eye((n-p),p);Q2=zeros((n-q),p); Q=[Q1;Q2]; % Step 4 F1=M1*A4-M1*A3*pinv(C1)*C2; F2=zeros((n-p),(n-p));F3=M3*A4-M3*A3*pinv(C2)*C1; F4=zeros((n-q),(n-p)); F=[F1 F2;F3 F4]; H1=M1*A3*pinv(C1); H2=M3*A3*pinv(C1); H=[H1;H2]; % Step 5 G=M*N*B; s=rand(1)+randn(1)*i; MAT=[s*eye(n)-A E;C zeros(2,2)]; rank(MAT) if rank(MAT)~=n+qdisp ('The method isn't useful'); end if $M^*N^*E \sim = [zeros((n-p),q); zeros((n-q),q)]$

disp ('First convergence condition isn't fulfilled'); end if $M*N*B-G\sim=zeros(n,m)$ disp ('Second convergence condition isn't fulfilled'); end if M*N*A-F*N~=H*C disp ('Third convergence condition isn't fulfilled'); end if P*N+Q*C~=I disp ('Fourth convergence condition isn't fulfilled'); end % K = 0K=zeros(1,n); sim ('schalgloor'); subplot(221); plot(t,x1,'b',t,xc1,'r--');grid; subplot(222); plot(t,x2,'b',t,xc2,'r--');grid; subplot(223); plot(t,x3,'b',t,xc3,'r--');grid; subplot(224); plot(t,x4,'b',t,xc4,'r--');grid; K=KK; h=figure;sim ('schalgloor'); subplot(221);plot(t,x1,'b',t,xc1,'r--');grid; subplot(222);plot(t,x2,'b',t,xc2,'r--');grid; subplot(223);plot(t,x3,'b',t,xc3,'r--');grid; subplot(224);plot(t,x4,'b',t,xc4,'r--');grid;

5.2 ALGLOOR program (rocket move)

% *Rocket move s*=10 run ALGX10; close all;clear C;clear contor;clear e; clear P1;clear P2;clear P3;clear P4; clear P;clear P;clear k1; KK=K: % Calculus data n=size(A,1);m=size(B,2); q=size(E,2); p=size(C,1);x0=[10;0;0;1]/100; xc0=[1;10;-1;2]/100;up=randn(2,1)% Matrices A and C partition A1=A(1:(n-p),1:(n-p)); A2=A(1:(n-p),(n-p+1):size(A,2));A3=A((n-p+1):size(A,1),1:(n-p)); A4=A((n-p+1):size(A,1),(n-p+1):size(A,2));C1=C(:,1:(n-p)); C2=C(:,(n-p+1):n);% Step 1 N1=zeros((n-q),(n-p)); N2=eye((n-q),p); N3=eye((n-p),(n-p));N4=zeros((n-p),p); N=[N1 N2;N3 N4]; % *Step* 2 M1=zeros((n-p),(n-q));M2=eye((n-p),(n-p));

M3=zeros((n-q),(n-q));M4=eye((n-q),(n-p));M=[M1 M2;M3 M4]; % Step 3 I1=eye((n-p),(n-p)); I2=zeros((n-p),(n-q));I3=zeros((n-q),(n-p));I4=eye((n-q),(n-q));I=[I1 I2;I3 I4]; P1=zeros((n-p),(n-q));P2=eye((n-p),(n-p));P3=zeros((n-q),(n-q));P4=-(pinv(C2))*C1;P=[P1 P2;P3 P4]; Q1=zeros((n-p),p); Q2=pinv(C2); Q=[Q1;Q2]; % Step 4 F1=zeros((n-p),(n-q));F2=A1;F3=zeros((n-q),(n-q));F4=A1;F=[F1 F2;F3 F4]; H1=A2;H2=A2; H=[H1;H2]; % Step 5 G=M*N*B: % Tests s=rand(1)+randn(1)*i; MAT=[s*eye(n)-A E;C zeros(2,2)]; if rank(MAT)~=n+qdisp ('The method isn't useful'); end if M*N*E~=[zeros((n-p),q);zeros((n-q),q)] disp ('First convergence condition isn't fulfilled'); end if $M*N*B-G\sim=zeros(n,m)$ disp ('Second convergence condition isn't fulfilled'); end if M*N*A-F*N~=H*C disp ('Third convergence condition isn't fulfilled'); end if P*N+O*C~=I disp ('Fourth convergence condition isn't fulfilled'); end % *K*=0 (*without controller*) K=zeros(1,n); sim('schrach'); subplot(221);plot(t,x1,'b',t,xc1,'r--');grid; subplot(222);plot(t,x2,'b',t,xc2,'r--');grid; subplot(223);plot(t,x3,'b',t,xc3,'r--');grid; subplot(224);plot(t,x4,'b',t,xc4,'r--');grid; h=figure; subplot(221);plot(t,e1);grid; subplot(222);plot(t,e2);grid; subplot(223);plot(t,e3);grid; subplot(224);plot(t,e4);grid; % K=K (with controller)

K=KK;h=figure;sim('schrach');subplot(221);plot(t,x1,'b',t,xc1,'r--');grid; subplot(222);plot(t,x2,'b',t,xc2,'r--');grid; subplot(223);plot(t,x3,'b',t,xc3,'r--');grid; subplot(224);plot(t,x4,'b',t,xc4,'r--');grid; h=figure; subplot(221);plot(t,e1);grid; subplot(222);plot(t,e2);grid; subplot(223);plot(t,e3);grid; subplot(224);plot(t,e4);grid;

5.3 ALGLX program

close all; A=[-0.007 0.012 -9.81 0; -0.128 -0.54 0 1; 0 0 0 1; 0.065 0.96 0 -0.99]; B=[0:-0.04;0;-12.5]; Q=[10 0 0;0 10 0;0 0 100 0;0 0 0]; R=[2]; [K,P,E] = LQR(A,B,Q,R); $I2=[1\ 0;0\ 1];$ N3=randn(4,3); contor=1; T(:,1)=B(:,1);for i=1:4 for j=1:3 T(i,j+1)=N3(i,j);end end Ab=(inv(T))*A*T;Bb=(inv(T)*B);Kb=place(Ab,Bb,E); e=eig(Ab-Bb*Kb); k1 = Kb(1); k21 = Kb(2);k22=Kb(3);k23=Kb(4); r1=5;Rb=[r1];R=Rb; Pb=r1*[k1 k21 k22 k23; k21 10 0; k22 0 1 0; k23 0 0 1]; ee=eig(Rb); Qb=-(Pb*Ab+(transpose(Ab))*Pb-Pb*Bb*Kb); PPP=transpose(inv(T))*Pb*inv(T); KKK=inv(R)*transpose(B)*PPP; EEE=eig(AB*KKK); m=rank(T); while real(EEE(1))>0 | real(EEE(2))>0 | real(EEE(3))>0 | real(EEE(4))>0 | m<4 N3=randn(4,3); contor=contor+1; T(:,1)=B(:,1)for i=1:4for j=1:3 T(i,j+1)=N3(i,j);end end Ab=(inv(T))*A*T; Bb=(inv(T)*B);

Kb=place(Ab,Bb,E); e=eig(Ab-Bb*Kb); k1=Kb(1);k21=Kb(2); k22=Kb(3);k23=Kb(4); r1=1;Rb=[r1];R=Rb; Pb=r1*[k1 k21 k22 k23; k21 10 0; k22 0 1 0; k23 0 0 1]; ee=eig(Rb); Qb=-(Pb*Ab+(transpose(Ab))*Pb-Pb*Bb*Kb); PPP=transpose(inv(T))*Pb*inv(T); KKK=inv(R)*transpose(B)*PPP; EEE=eig(AB*KKK); m=rank(T);end Q=transpose(inv(T))*Qb*inv(T); R=Rb; [KK, PP, EE] = LQR(A, B, Q, R);

5.4 ALGLX10 program

close all;clear all; d1=1.5; d2=40; d3=-20; d4=1.2; V=400; T1=1/d1;Tv=T1; w0=sqrt(d1*d4-d3); csi=(d1+d4)/(2*w0);kd=d2; $a0=w0^2-(2*csi*w0)/Tv+1/(Tv^2);$ a1=2*csi*w0-1/Tv;k0=70; A=[0 1 0 0;0 0 V/Tv 0;0 0 -1/Tv 1;0 0 -a0 -a1]; B = [0;0;0;kd]; $C = [0 \ 0 \ 1 \ 0; 0 \ 0 \ 0 \ 1];$ $E = [0 \ 0; 0 \ 0; 1 \ 0; 0 \ 1];$ $Q = [10 \ 0 \ 0; 0 \ 10 \ 0; 0; 0 \ 100 \ 0; 0 \ 0 \ 1]; R = [2];$ [K,P,E] = LQR(A,B,Q,R); I2=[1 0;0 1];% Matrix T definition N3=randn(4,3); contor=1; T(:,1)=B(:,1); **for** i=1:4 for j=1:3T(i,j+1)=N3(i,j);end end Ab=(inv(T))*A*T;Bb=(inv(T)*B);Kb=place(Ab,Bb,E); e=eig(Ab-Bb*Kb); k1 = Kb(1); k21 = Kb(2);k22=Kb(3);k23=Kb(4); r1=1;Rb=[r1];R=Rb; Pb=r1*[k1 k21 k22 k23;k21 1 0 0; k22 0 1 0;k23 0 0 1]; ee=eig(Rb); Qb=-(Pb*Ab+(transpose(Ab))*Pb-Pb*Bb*Kb); PPP=transpose(inv(T))*Pb*inv(T); KKK=inv(R)*transpose(B)*PPP; EEE=eig(A-B*KKK);m=rank(T); while real(EEE(1))>0 | real(EEE(2))>0 | real(EEE(3))>0 | real(EEE(4))>0 | m<4 % Matrix T definition

```
N3=randn(4,3);
contor=contor+1;
T(:,1)=B(:,1);
for i=1:4
  for j=1:3
    T(i,j+1)=N3(i,j);
  end
end
Ab=(inv(T))*A*T;
Bb=(inv(T)*B);
Kb=place(Ab,Bb,E);
e=eig(Ab-Bb*Kb);
k1=Kb(1);k21=Kb(2);k22=Kb(3);k23=Kb(4);
r1=1;Rb=[r1];R=Rb;
Pb=r1*[k1 k21 k22 k23;k21 1 0 0;
       k22 0 1 0;k23 0 0 1];
ee=eig(Rb);
Qb=-(Pb*Ab+(transpose(Ab))*Pb-Pb*Bb*Kb);
PPP=transpose(inv(T))*Pb*inv(T);
KKK=inv(R)*transpose(B)*PPP;
EEE=eig(A-B*KKK);m=rank(T);
end
% Variant 1
O=transpose(inv(T))*Ob*inv(T);R=Rb;
[KK, PP, EE] = LQR(A, B, Q, R);
```

6 Conclusion

A new algorithm for project of a reduced order observer is presented; it may be used in different applications and in case of aircrafts' automat command system. Theoretical results are validated by numerical simulations using calculus program made by authors using dynamic models of aircrafts, rockets' movement equations and optimal control laws.

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