Performance Analysis of Spectrum Sensing for RF Receiver Structure in Cognitive Radio Networks

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Abstract—This paper analyses the performance of spectrum sensing in terms of the throughput of a cognitive radio (CR) system. Dealing with the optimization problem of spectrum sensing, this paper evaluates the throughput of a CR system by considering such situations as the penalty time of a channel search and incumbent user (IU) detection delay caused by a missed detection of an incumbent signal. Also, this paper suggests a serial channel search scheme as the search method for a vacant channel, and derives its mean channel search time by considering the penalty time due to a vacant channel search error. The numerical results suggest the optimum sensing time of the channel search process using the derived mean channel search time of a serial channel search in the case of a sensing hardware structure with single radio frequency (RF) path. It also demonstrates that the average throughput is improved by two separate RF paths in spite of the hardware complexity of an RF receiver.

I. Introduction

The licensed spectrum allocation method appears to have low spectrum utilization in many parts of the frequency band. In the Federal Communications Commission (FCC) definition, a cognitive radio (CR) system is a spectrum utilization that allows CR users to use spectrum licensed to incumbent users when they are inactive [1]-[3]. In order to utilize the inactive frequency bands, CR users must perform the spectrum sensing for the CR-used frequency bands frequently or periodically. The main reasons for spectrum sensing in CR users can be classified into two types. The first is to gather vacant channels for CR users, and the second is to check the appearance of incumbent users for the corresponding channel that a CR user is utilizing. This paper evaluates the sensing performance in terms of throughput efficiency by considering channel monitoring, a channel search, and a channel switch. The tradeoff between sensing time and throughput in a CR system is an important issue for sensing optimization. Because there is a tradeoff between channel monitoring time and throughput, researches on the optimal sensing time have been performed in terms of throughput or throughput efficiency [4], [5].

II. Performance Analysis of Spectrum Sensing

In order to use the license band of incumbent users, CR users must investigate the vacancy of frequency channels using spectrum sensing.

Figure 1 shows the procedure of vacant channel utilization and periodic spectrum sensing in a CR system. In the figure, the channel availability check duration, $T_{\text{check}}$, is the time during which a CR system searches vacant channels for CR network initialization. When a CR system completes a search of vacant channels initially, the CR system sets up a CR-operating channel by selecting one of the vacant channels for data communication. The CR users perform the spectrum sensing to identify the appearance of an incumbent user through a channel monitoring process after a channel is setup, as shown in Fig. 1.
The channel monitoring process to identify the appearance of incumbent users is performed periodically using sensing parameters such as sensing period, $T_{\text{frame}}$, and sensing time, $T_{\text{monitor}}$, as shown in Fig. 1. Then, if an incumbent signal appears in the frequency channel corresponding to the CR-operating channel, the CR user must switch to another vacant channel to be assigned as a backup channel. Here, the backup channel is defined as a vacant channel that can be assigned as a CR-operating channel during the channel switch process. This paper evaluates the sensing performance by considering the channel search, channel switch, and channel monitoring processes, as shown in Fig. 1.

Also, for the appearance of an incumbent signal, the conventional research analyzes the sensing performance for cases in which a CR system stops its CR operation and searches other vacant channels. In this paper, however, owing to the sensing structure of two separate RF paths as shown in Fig. 2, the channel switch time shortens by searching vacant channels during data transmission time, $T_{\text{data}}$.

In Fig. 2, since the CR terminal and CR base station do not know the appearance position of an incumbent user, a spectrum sensing antenna is used as an omni-directional antenna. However, in order to reduce the interference of signals other than those from the CR system, the data antenna for CR communications is composed of a directional antenna. Thus, in order to obtain the directional antenna gain of the data antenna, the sensing structure with two separate RF paths has an advantage in terms of the sensing antenna and data antenna configuration, as shown in Fig. 2.

For cases when an incumbent signal appears during a CR operation, the previous research proposes a method in which the CR user searches another vacant channel after determining the appearance of the incumbent signal [4]. A state diagram of the serial channel search scheme for a serial channel search performed on N frequency channels, except for the $k$-th CR-operating channel, when the number of total frequency channels in the CR system is given as N+1, is shown in Fig. 3.

![State diagram of the serial channel search process.](Image)

In Fig. 3, a search success (SS) state is defined as the state that acquires a vacant channel successfully, while a missed detection (MD) state is described as a case in which a vacant channel is declared incorrectly due to a decision error on the existence of an incumbent signal when such an incumbent signal exists. From Fig. 3, $C_i(i=0,1,\ldots,N-1)$ is the $i$-th frequency channel, $h_i(z)$ is the path function that acquires the vacant channel successfully, and $h_{md}(z)$ is the path function that detects correctly the existence of an incumbent signal. Finally $h_{md}(z)$ is the path function that declares incorrectly a vacant channel during the existence of an incumbent signal, and $z^p$ is the penalty time that is given by an incorrect decision of a vacant channel due to a missed detection of an incumbent signal in the channel search process.

From Fig. 3, $h_i(z)$, $h_j(z)$, and $h_{md}(z)$ are represented by

$$h_i(z) = P(H_i)(1-P'_i)z, \quad (1)$$

$$h_j(z) = \left\{P(H_i)P'_i + P(H_i)P'_j\right\}z, \quad (2)$$

and

$$h_{md}(z) = P(H_i)(1-P'_j)z, \quad (3)$$

where $P(H_i)$ is defined as the probability that an incumbent signal is inactive, and $P(H_i)$ is defined as the probability that an incumbent signal is active, $P(H_i) + P(H_i) = 1 [5]$. $P'_i$ and $P'_j$ are the probabilities of a signal detection and false alarm in the channel search process, respectively. From Fig. 3, if $h_i(z)$ is the success function of the channel search of a vacant channel and $h_{mv}(z)$ is the failure function of the
channel search of a vacant channel, \( h_v(z) \) and \( h_m(z) \) are given by
\[
\begin{align*}
    h_v(z) &= h_s(z) + h_m(z) z^p \\
    &= P(H_0)P_s^c + P(H_1)P_s^f z + P(H_1)(1-P_s^f)z^{p+1}.
\end{align*}
\] (4)

In Fig. 3, performing a serial channel search for \( N \) available channels, the success function of the channel search, \( H_s(z) \) is given by
\[
H_s(z) = h_s(z) + h_v(z) h_s(z) + h_m(z) h_s(z) + \ldots + h_m^{N-1}(z) h_s(z) = h_s(z) \sum_{m=0}^{N-1} h_v(z).
\] (5)

The transfer function that starts a channel search at the \( i \)-th channel is given by
\[
U_i(z) = \frac{H_s(z)}{1-h_v(z)}.
\] (6)

From Fig. 3, since all channels are equally likely a priori, the total transfer function averaged over all \( N \) frequency channels is represented by
\[
U(z) = \frac{1}{N} \sum_{i=1}^{N} U_i(z) = \frac{H_s(z)}{1-h_v(z)}.
\] (7)

Thus, the mean channel search time of the serial channel search scheme is given by
\[
\bar{T}_{\text{search}} = \frac{d}{dz} \ln U(z) \bigg|_{z=1} T_{\text{search}} = \left[ \frac{H_s'(1) + Nh_v'(1)h_m^{N-1}(1)}{H_s'(1) - 1-h_v'(1)} \right] T_{\text{search}},
\] (8)

where \( T_{\text{search}} \) is the sensing time per one channel in the channel search process, \( H'(l) = H(z) \bigg|_{l=1} \), \( h'(l) = h(z) \bigg|_{l=1} \),
\[
H'(l) = \frac{d}{dz} H(z) \bigg|_{z=l}, \text{ and } h'(l) = \frac{d}{dz} h(z) \bigg|_{z=l}.
\]

The sensing structure of a single RF path for sensing and CR operation performs both the channel search and channel switch processes during the appearance of incumbent users, while the sensing structure of two separate RF paths conducts only the channel switch process due to the acquirement of vacant channels using the data transmission time. Thus, the channel movement time for the sensing structure with two separate RF paths is given by
\[
\bar{T}_{\text{move}} = T_{\text{frame}} + T_{\text{switch}}.
\] (9)

And the channel movement time for the sensing structure with a single RF path is given by
\[
\bar{T}_{\text{move}} = T_{\text{frame}} + \bar{T}_{\text{search}} + T_{\text{switch}}.
\] (10)

In general, the throughput is defined as a ratio of the data transmission time divided by the total channel occupation time of the CR-operating channel. From Fig. 2, the total channel occupation time of the CR-operating channel, \( \bar{T}_{\text{total}} \) is given by
\[
\bar{T}_{\text{total}} = P(H_0)(1-P_s^c)T_{\text{frame}} + P(H_s)P_s^c \bar{T}_{\text{move}} + P(H_1)(1-P_s^f) \bar{T}_{\text{delay}},
\] (11)

where \( P_s^c \) and \( P_s^f \) are the probability of signal detection and the probability of a false alarm in the channel monitoring process, respectively. Also, \( T_{\text{delay}} \) is the IU detection delay, which is defined as a transmission delay caused by a transmission error due to the interference of an incumbent signal when detection of the incumbent signal is missed. \( T_{\text{delay}} \) can be described as a transmission delay due to a missed detection for continuous channel monitoring when a transmission error occurs at the CR-operating channel. The detection of the incumbent signal can also be determined by a rapid increase of bit error rate measured by \( M \) data frames when a missed detection of an incumbent signal occurs during the continuous channel monitoring process. In such a case, \( T_{\text{delay}} \) is represented as \( MT_{\text{frame}} \).

In this paper, the derivation of throughput considering the serial channel search process and IU detection delay is as follows. The mean data transmission time of the CR-operating channel is given by
\[
\bar{T}_{\text{data}} = \bar{T}_{\text{frame}} \left[ P(H_0)(1-P_s^c) + P(H_s)P_s^c \right],
\] (12)

In (18), the throughput for the appearance of an incumbent signal is almost zero due to the transmission error caused by the interference of the incumbent signal. Thus, the mean data transmission time can be represented by
\[
\bar{T}_{\text{data}} = \bar{T}_{\text{data}} \left[ P(H_0)(1-P_s^c) + P(H_s)(1-P_s^f) \right].
\] (13)

Finally, the average throughput using (17) and (19) can be represented by
where $\eta$ is the data rate of the CR system.

### III. Numerical Results

In this paper, an incumbent signal for the numerical results is assumed to be a binary phase-shift keying signal with a 10 kHz bandwidth. Energy detection is selected for the incumbent signal, and the sampling rate of the energy detector is assumed to be 10 kHz. The channel model of both the incumbent and CR systems is assumed to be a Rayleigh fading channel. Considering the Rayleigh fading channel [7], [8], the probabilities of signal detection and a false alarm during the channel monitoring and channel search are given by

$$
P^j_s = e^{-\lambda/V_F} \sum_{k=0}^{\infty} \frac{(\lambda/V_F)^k}{k!}, i = s, m,
$$

$$
P^j_f = e^{-\lambda/V_N} \sum_{k=0}^{\infty} \frac{(\lambda/V_N)^k}{k!}, i = s, m,
$$

where $L$ is the time-bandwidth product, $B_T$ the channel bandwidth, $B$ is multiplied by the sensing time $T$, and $\lambda$ and $V_N$ are the detection threshold and noise variance, respectively. $V_F$ is given by

$$
V_F = V_N(1 + \mu),
$$

where $\mu$ is the average SNR of the incumbent signal for the Rayleigh fading channel.

The number of the frequency channels is assumed to be 31 with a bandwidth of 10 kHz in the operating bandwidth of the CR system. The frame length of the CR system, $T_{frame}$, is assumed to be 100 ms, and the channel switch time, $T_{switch}$, is then assumed to 10 ms for the control signal transmission and channel switch.

Figure 4 shows the mean channel search time for the time-bandwidth product.

Figure 5 shows the average throughput for the channel monitoring time according to the sensing hardware structure and changes of channel search time. In Fig. 5, the solid lines represent the results for the sensing hardware related to spectrum sensing and data communication in the case of two separate RF paths, while the dotted lines are those for a single RF path. Actually, the average throughput for the channel monitoring time according to the sensing hardware structure...
has a large difference when the sensing time of the channel monitoring and channel search enlarges.

Figure 6 shows the average throughput for the channel monitoring time according to the IU detection delay. In Fig. 6, an IU detection delay is defined as a transmission delay caused by a transmission error due to interference of an incumbent signal when detection of the incumbent signal is missed. From (20), if the average throughput is evaluated with no IU detection delay, as shown in Fig. 6, an analytical error is caused as a result of the throughput improvement in cases of shorter channel monitoring times. In Fig. 6, the changes of average throughput for the IU detection delay are very small when the channel monitoring time is assigned as 5ms or more.

IV. Conclusion

This paper handles the optimization of spectrum sensing in terms of CR system throughput. This paper also evaluates the throughput of a CR system by considering such situations as channel search, channel switch, and sensing structure, and derives the mean channel search time of a serial channel search by considering the penalty time due to a search error. The numerical results suggest an optimum sensing time of the channel search process using the derived mean channel search time of a serial channel search. It also demonstrates that the average throughput is improved by two separate RF paths in spite of the hardware complexity of the RF receiver.

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V. References