# A Pipelined Divider with a Small Lookup Table 

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#### Abstract

The design of fast dividers is an important issue in high speed computing because division account for a significant fraction of the total arithmetic operation. Taylor series expansion is a well-known multiplicative scheme for high-performance division implementation. This study presents a simple architecture that implements a pipelined divider including the first 6 terms of the Taylor series expansion for approximation. Results show that the developed pipelined divider takes a lookup table of 32B for single precision with a latency of 8.90 ns , and 56 KB for double precision with 11.46 ns , where the circuit is synthesized with TSMC $0.18 \mu \mathrm{~m}$ digital CMOS standard cell library.


Key-words: Pipelined Divider, Lookup Table, Taylor Series Expansion, Approximation, Newton-Raphson Algorithm.

## 1. Introduction

The design of fast dividers is an important issue in high speed computing because division account for a significant fraction of the total arithmetic operation [1]. Most implementations for the division are based on the high-radix SRT algorithm that uses a recurrence producing one quotient digit for each step [2-9]. However, high-performance implementations tend to use multiplier-based convergence schemes, referred to as Multiplicative schemes. Basically, a division ( $\mathrm{X} / \mathrm{Y}$ ) is equivalent to a multiplication of $X$ and $(1 / Y)$, the reciprocal of the divisor Y. Thus, a multiplier and lookup table(s) for approximating ( $1 / \mathrm{Y}$ ) are employed in the multiplicative scheme.

Newton-Raphson and series-expansion algorithms are well-known multiplicative schemes that may require a large lookup table for high-speed operation. For example, the lookup table for a 16 -bit seed Newton-Raphson or series-expansion divider is 64 KB (bytes) [10]. These 16 -bit seed dividers execute a division operation with two iterations in single precision. If 8 -bit seed dividers are used, the table size is smaller as 128 B , but their latency is longer because they require three iterations. An accurate quotient approximation, which used Taylor-
series expansion, has two lookup tables, requiring 400B for single precision [11].

A high-radix pipelinable-division algorithm based on the well-know Taylor series expansion was proposed [12]. The algorithm took the first two terms of the Taylor series expansion for approximation. As illustrated in Figure 1(a), the algorithm provided a simple architecture using a lookup table of 13 KB for single precision. It is virtually impossible to implement the divider in double precision because a huge lookup table, nearly 470 MB , is required. Therefore, the size of the lookup table is a critical factor to implement a pipelinable divider. Recently, a cost-effective pipelined divider, as shown in Figure 1(b), with a small lookup table was proposed to reduce the table size for implementing in double precision [13]. The algorithm took the first four terms of Taylor series expansion for approximation. Results show that table size can be reduced from 13KB to 208B for single precision, and from 470 MB to 56 KB for double precision.

Theoretically speaking, including more terms of Taylor series expansion for approximation will improve the accuracy of approximation and thus reducing the size of lookup table. However, including more terms also increases more real-time computations and thus requiring more hardware and longer
latency. The question that naturally arises is whether there exists a point where the hardware saving due to the table size reduction is offset the increased hardware due to the need of more real-time computations.

(a)


Figure 1. Block Diagrams for Division Algorithms in Single Precision: (a) [12]; and (b) [13].

This study develops a pipelined division algorithm that includes six terms of Taylor series expansion for approximation. Results will show that the table size can be further reduced from 208B in [13] to 36B for single precision, and from 56 KB to 1.28 K for double precision. The area improvement is achieved at the cost of longer latency, where the latency is increased from 7.18 ns to 8.90 ns for single precision, from 9.46 ns to 11.46 ns for double precision. This study also shows that the structure including 6 terms for approximation provides the minimal hardware with a reasonable latency.

In the next section, the basic division algorithm in [13] is reviewed. Section 3 presents the proposed pipelined division algorithms and their hardware implementation. The designs of the division algorithms with four-terms, six terms, and eight terms
of Taylor series expansion and their structures are respectively also developed to discuss the design trade-offs. The total error is the errors accumulated at each step of a division algorithm, and it determines the size of lookup table and bit-sizes of the multipliers. The error analysis of each developed algorithm will also presented in this section. Section 4 presents the experimental results. Finally, a concluding remark is given in Section 5.

## 2. Basic Division Algorithms

Let $X$ and $Y$ be two $m$-bit normalized fixed-point radicand, where $\mathrm{X}=1 . \mathrm{x}_{1} \mathrm{x}_{2} . . \mathrm{x}_{\mathrm{m}}$ and $\mathrm{Y}=1 . \mathrm{y}_{1} \mathrm{y}_{2} . . \mathrm{y}_{\mathrm{m}}$. To calculate $\mathrm{X} / \mathrm{Y}, \mathrm{Y}$ is first decomposed into two groups: the higher order bits $\left(\mathrm{Y}_{\mathrm{h}}\right)$ and the lower order bits $\left(\mathrm{Y}_{1}\right)$, i.e., $\mathrm{Y}=\mathrm{Y}_{\mathrm{h}}+\mathrm{Y}_{1}$, where $\mathrm{Y}_{\mathrm{h}}=$ $1 . y_{1} y_{2} . . y_{p-1}$ and $\mathrm{Y}_{1}=0.00 . .0 \mathrm{y}_{\mathrm{p}} \mathrm{y}_{\mathrm{p}+1} . . \mathrm{y}_{\mathrm{m}}$. Therefore, $\mathrm{Y}_{\mathrm{h}} \gg \mathrm{Y}_{1}$ and $\mathrm{Y}_{1} / \mathrm{Y}_{\mathrm{h}}$ is approximately zero. In other words,

$$
\begin{gather*}
1 \leq \mathrm{X}, \mathrm{Y}<2 ; 1 \leq \mathrm{Y}_{\mathrm{h}}<2-2^{-\mathrm{p}+1} \\
\text { and } 0 \leq \mathrm{Y}_{1}<2^{-\mathrm{p}+1} \tag{1}
\end{gather*}
$$

A division operation can be represented by Taylor series expansion as follows:

$$
\begin{align*}
\frac{X}{Y} & =\frac{X}{Y_{h}+Y_{l}} \\
& =\frac{X}{Y_{h}}\left[1-\frac{Y_{1}}{Y_{h}}+\left(\frac{Y_{I}}{Y_{h}}\right)^{2}-\ldots-\left(\frac{Y_{I}}{Y_{h}}\right)^{2 n-1}+\ldots\right] \tag{2}
\end{align*}
$$

If we take the first two terms of (2) to approximate the division (X/Y), we obtain

$$
\begin{equation*}
\frac{X}{Y} \approx \frac{X\left(Y_{h}-Y_{l}\right)}{Y_{h}^{2}}=X\left(Y_{h}-Y_{l}\right)\left(\frac{1}{Y_{h}{ }^{2}}\right) \tag{3}
\end{equation*}
$$

Therefore, a division operation is executed by multiplying $\mathrm{X},\left(\mathrm{Y}_{\mathrm{h}}-\mathrm{Y}_{1}\right)$, and $\left(1 / \mathrm{Y}_{\mathrm{h}}{ }^{2}\right)$, where $\left(1 / \mathrm{Y}_{\mathrm{h}}{ }^{2}\right)$ is approximated by a lookup table, and the multiplication of $\mathrm{Y}\left(\mathrm{Y}_{\mathrm{h}}-\mathrm{Y}_{1}\right)$ is computed by a Booth multiplier [12].

The division operation is simplified by first defining a coarse quotient $Q^{\prime}=X A$, where $A=\left(Y_{h^{-}}\right.$ $\left.\mathrm{Y}_{1}\right) / \mathrm{Y}_{\mathrm{h}}{ }^{2}$, and the subdivident $\mathrm{X}^{\prime}=\mathrm{X}-\mathrm{YQ}^{\prime}$. Let $\mathrm{Q}^{\prime \prime}=$ $X^{\prime}\left(Y_{h}-Y_{1}\right)\left(1 / Y_{h}{ }^{2}\right)=X^{\prime} A$. The quotient is calculated as

$$
\begin{align*}
X / Y & \approx Q^{\prime}+Q^{\prime \prime}=\left(X+X^{\prime}\right) A=\left(2 X-Y Q^{\prime}\right) A \\
& =(A X)(2-A Y) \tag{4}
\end{align*}
$$

where both (AY) and (AX) are calculated through parallel multiplications, as shown in Figure 1(b), and (2-AY) is a two's complement of (AY). The
approximating error, i.e., the difference between $(\mathrm{X} / \mathrm{Y})$ and ( $\left.\mathrm{Q}^{\prime}+\mathrm{Q}^{\prime \prime}\right)$, is expressed as

$$
\begin{gather*}
\Delta=\frac{X}{Y}-(2-A Y) A X, \text { and } \\
\Delta=X\left\{\frac{1}{Y}-\left[2-\frac{Y\left(Y_{h}-Y_{l}\right)}{Y_{h}^{2}}\right]\left[\frac{Y_{h}-Y_{l}}{Y_{h}^{2}}\right]\right\}=\frac{X}{Y}\left(\frac{Y_{l}}{Y_{h}}\right)^{4} \tag{5}
\end{gather*}
$$

The size of lookup table and the bit-widths of the multipliers are determined by the possible errors induced. Four possible errors were considered [13]: (a) Error caused by the restriction in the number of entries in the lookup table; (b) Error caused by the bit-width restriction of the lookup table; (c) Error caused by the rounding positions; and (d) Error caused by the bit-inversion.

The first error is calculated by subtracting the actual quotient from the ideal quotient, where the error is expressed as in (5). The number of entries and output bit-width in the lookup table are $2^{(p-1)}$ and q , respectively. Based on the error analysis [13], both $p$ and $q$ are selected from the following inequalities,

$$
\begin{equation*}
\mathrm{p}>(\mathrm{m}+4) / 4 \text { and } \mathrm{q} \geq-2 \mathrm{p}+\mathrm{m}+5 \tag{6}
\end{equation*}
$$

For single precision with $\mathrm{m}=24$, by (6), $\mathrm{p}=8$ and $\mathrm{q}=13$ are resulted. The table size is $2^{\mathrm{p}-1} * \mathrm{q}=1664$ bits or 208 Bytes, and four multipliers with the dimensions of $24 \times 13,24 \times 15,24 \times 15$, and $28 \times 28$ are employed with a latency of 11.01 ns , where the circuit was synthesized with Samsung MDL110 $0.25 \mu \mathrm{~m}$ CMOS standard cell library [14]. Note that the division algorithm in Figure 1(a) requires a lookup table with the size of 13 KB and two multipliers with the dimensions of $26 \times 26$ and $24 \times 24$, and takes a latency of 7.62 ns . On the other hand, for double precision with $\mathrm{m}=53$, by (6), $\mathrm{p}=15$ and $\mathrm{q}=28$ are concluded, where the table size is 448 K bits. In addition, the division algorithm requires three $53 \times 28$ multipliers and one $58 \times 58$ multiplier are employed and takes a latency of 24.2 ns [13].

Unfortunately, the latency of 24.2 ns for double precision is impractical. The longer latency is due to the use of table size of 56 KB . Thus, it motivates this study to develop a simple, yet better latency for the pipelined division in double precision.

## 3. Developed Division Algorithms

This section presents the developed pipelined division algorithms.

### 3.1. Division Algorithms and Hardware Structures

Consider the approximating value $\mathrm{Q}_{2 n}$ that includes the first 2 n terms of the Taylor series expansion in (2), i.e.,

$$
\begin{align*}
\mathrm{Q}_{2 \mathrm{n}} & =\frac{X}{Y_{h}}\left[1-\frac{Y_{l}}{Y_{h}}+\left(\frac{Y_{I}}{Y_{h}}\right)^{2}-\left(\frac{Y_{J}}{Y_{h}}\right)^{3}+\ldots-\left(\frac{Y_{J}}{Y_{h}}\right)^{2 n-1}\right] \\
& =\frac{X}{Y_{h}}\left[1-\frac{Y_{l}}{Y_{h}}\right]\left[1+\left(\frac{Y_{J}}{Y_{h}}\right)^{2}+\left(\frac{Y_{l}}{Y_{h}}\right)^{4}+\ldots+\left(\frac{Y_{l}}{Y_{h}}\right)^{2 n-1}\right] \\
& =A X\left[1+\left(\frac{Y_{I}}{Y_{h}}\right)^{2}+\left(\frac{Y_{J}}{Y_{h}}\right)^{4}+\ldots+\left(\frac{Y_{l}}{Y_{h}}\right)^{2 n-1}\right] \tag{7}
\end{align*}
$$

The approximating value can be expressed as in the following theorem,

## Theorem 1.

The approximating value $\mathrm{Q}_{2 \mathrm{n}}$ can be expressed as

$$
\begin{equation*}
\mathrm{Q}_{2 \mathrm{n}}=\mathrm{Q}_{2(\mathrm{n}-1)}(1-\mathrm{AY})+\mathrm{Q}_{2} \tag{8}
\end{equation*}
$$

where $Q_{2}=X A$ and $A=\left(Y_{h}-Y_{1}\right) / Y_{h}{ }^{2}$.

The approximating value $\mathrm{Q}_{2 \mathrm{n}}$ can also be expressed as

$$
\begin{align*}
\mathrm{Q}_{2 \mathrm{n}} & =A X\left[\frac{1-\left(Y_{l} / Y_{h}\right)^{2 n}}{1-\left(Y_{l} / Y_{h}\right)^{2}}\right]=A X\left[\frac{1-\left(Y_{l} / Y_{h}\right)^{2 n}}{A Y}\right] \\
& =\left(\frac{X}{Y}\right)\left[1-\left(Y_{l} / Y_{h}\right)^{2 n}\right] \tag{9}
\end{align*}
$$

Therefore, the difference between the ideal quotient and actual quotient, or approximating error, is

$$
\begin{equation*}
(\mathrm{X} / \mathrm{Y})-\mathrm{Q}_{2 \mathrm{n}}=\left(\frac{X}{Y}\right)\left(\frac{Y_{l}}{Y_{h}}\right)^{2 n} \tag{10}
\end{equation*}
$$

By (9) \& (10), the approximating value and error for $\mathrm{n}=2$ are respectively

$$
\begin{align*}
Q_{4}= & Q_{2}(1-A Y)+Q_{2}=Q_{2}(2-A Y)=A X(2-A Y)  \tag{11}\\
& (X / Y)-Q_{4}=(X / Y)\left(Y_{1} / Y_{h}\right)^{4} \tag{12}
\end{align*}
$$

In other words, the complicated derivations in (4) and (5) can be simply derived from the Taylor series expansion as shown in (11) and (12).

Figure 2 shows an alternative pipelined divider implementing Equation (11). Instead of using a bitinversion process to find (2-AY) in Figure 1(b), this algorithm calculates the final quotient by adding $\mathrm{Q}_{2}=\mathrm{XA}$ to the product of XA and (1-AY), where a simple bit complementer is employed to derive (1-

AY). In practice, the above multiplication and addition may be performed by a MAC (Multiplying and ACcumulating) unit. Both structures in Figures 2 and 1 (b) should have the same hardware cost and latency.

Note that the accuracy of the approximating value can be improved by including more terms in (7) for approximation at the cost of more hardware and longer latency for calculating the additional terms. On the other hand, higher accuracy of the approximating value requires smaller size of lookup table. In this study, the table size, total hardware cost, and latency of pipelined division algorithms with $\mathrm{n}=2,3$, and 4 , are presented and evaluated


Figure 2. Developed Pipelined Division Algorithm with $\mathrm{n}=2$ in Single Precision:

By theorem 1, the approximating value and error for $\mathrm{n}=3$ are respectively

$$
\begin{align*}
& \mathrm{Q}_{6}=\mathrm{Q}_{4}(1-\mathrm{AY})+\mathrm{Q}_{2}  \tag{13a}\\
& (\mathrm{X} / \mathrm{Y})-\mathrm{Q}_{6}=(\mathrm{X} / \mathrm{Y})\left(\mathrm{Y}_{1} / \mathrm{Y}_{\mathrm{h}}\right)^{6} \tag{13b}
\end{align*}
$$

and those for $\mathrm{n}=4$ are

$$
\begin{align*}
& \mathrm{Q}_{8}=\mathrm{Q}_{6}(1-\mathrm{YA})+\mathrm{Q}_{2}  \tag{14a}\\
& (\mathrm{X} / \mathrm{Y})-\mathrm{Q}_{8}=(\mathrm{X} / \mathrm{Y})\left(\mathrm{Y}_{1} / \mathrm{Y}_{\mathrm{h}}\right)^{8} \tag{14b}
\end{align*}
$$

For $\mathrm{n}=3$, an additional MAC unit is required, while two MAC units are needed for $n=4$.

Similar to the error analysis presented in [13], both p and q are determined by the following inequalities,

$$
\begin{equation*}
2^{-4 p+5}<2^{-m+1}, \text { or } \quad \mathrm{p}>(\mathrm{m}+4) / 4 \tag{15a}
\end{equation*}
$$

$$
\begin{align*}
& 2^{-2 p-q+6} \leq 2^{-m+1}, \text { or } \quad q \geq-2 p+m+5  \tag{15b}\\
& 2^{-2 p-m 1+5} \leq 2^{-3} * 2^{-m+1}, \text { or } m_{1} \geq-2 p+m+7  \tag{15c}\\
& 2^{-m 3+2} \leq 2^{-3} * 2^{-m+1}, \text { or } m_{2}=m_{3} \geq m+4 \tag{15d}
\end{align*}
$$

Thus, the dimensions of the four multipliers are $m$ $\mathrm{xq}, \mathrm{mx} \mathrm{m}_{1}, \mathrm{mx} \mathrm{m}_{1}$, and $\mathrm{m}_{2} \mathrm{xm}_{2}$, respectively.

For single precision with $m=24$, by (15a) and ( 15 b ), $\mathrm{p}=8$ and $\mathrm{q}=13$ are selected. $\mathrm{By}(15 \mathrm{c}), \mathrm{m}_{1}=15$, and by $(15 d), m_{2}=m_{3}=28 . m_{4}=24$ is the bit-width of the final quotient. Therefore, the dimensions of the multipliers are $24 \times 13,24 \times 15,24 \times 15$, and $28 \times 28$. On the other hand, for double precision with $\mathrm{m}=53$, $\mathrm{p}=15$ and $\mathrm{q}=28$ are resulted. By (15c), $\mathrm{m}_{1}=30$, and by ( 15 d ), $\mathrm{m}_{3}=\mathrm{m}_{2}=57$. Thus, the dimensions of the multipliers are $53 \times 28,53 \times 30,53 \times 30$, and $57 \times 57$.

### 3.2. Experimental Results

The pipelined division algorithms in Figures 2have been developed, where the circuits were synthesized with TSMC 0.18 mm digital CMOS standard cell library. The pipelined divider with $\mathrm{n}=2$ for single precision requires a lookup table with a size of $2^{7} \times 13$ or 208 B which takes an area of 63,000 $\mu \mathrm{m}^{2}$ and a delay of 1.21 ns , as shown in Table 2(a). The dimensions of four multipliers in this divider are $24 \times 13,24 \times 15,24 \times 15$, and $28 \times 28$, respectively, and their corresponding areas are $33,277 \mu^{2}$, $41,014 \mu^{2}, 41,014 \mu^{2}$, and $86,619 \mu \mathrm{~m}^{2}$. This concludes that the total area of the pipelined divider with $n=2$ for single precision is $264,924 \mu \mathrm{~m}^{2}$. The critical path of the divider includes the lookup table, $\mathrm{M}_{1}, \mathrm{M}_{3}$, and $\mathrm{M}_{4}$, and their corresponding delays are $1.21 \mathrm{~ns}, 1.81 \mathrm{~ns}, 2.28 \mathrm{~ns}$, and 2.88 ns , respectively. Thus, the divider has a delay of 7.18 ns . For double precision, the pipelined divider with $\mathrm{n}=2$, the size of lookup table is $2^{14} \times 28$, or 56 KB and takes $207,02,938 \mu^{2}$ in area and 1.67 ns in delay. The total area of the divider is $21,422,752$ $\mu \mathrm{m}^{2}$ with a delay of 9.46 ns .

## 4. Conclusions

Taylor series expansion is a well-known multiplicative scheme for high-performance division implementation. A high-radix pipelinable division algorithm based on Taylor series expansion [12] included the first two terms of Taylor series expansion, or $\mathrm{n}=1$, for approximation. It provides a simple architecture with a lookup table of 13 KB for single precision. The cost-effective pipelined divider [13] with the first 4 terms of Taylor series
expansion, reduced the table size from 13 KB to 208B for single precision, and from 470MB to 56 KB for double precision. This paper presents a simple architecture that implements a pipelined divider with the first 6 terms of the Taylor series expansion. Results show that the developed pipelined divider further reduces the size of lookup table from 208B to 32B for single precision, and from 56 KB to 1.28 KB for double precision. This study also shows that the table size can be further reduced as more terms in Taylor series expansion are included for approximation. However, including more terms require additional hardware for calculating the extra terms. As a result, the pipelined divider with the first 6 terms provides an optimal solution which requires the minimum area with a reasonable latency.

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