Design and Fabrication of High-Q Spiral Inductors Using MEMS Technology

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Abstract: This paper presents the design and modeling, fabrication and characterization of suspended spiral inductors on silicon substrate. The substrate materials underneath the inductor coil are removed by micromachining process to reduce the substrate loss which enables high frequency operation. A complete Library of spiral inductors with different line width, line spacing and number of turns have been designed and simulated. The results show that the Q factor as well as the self resonance frequency \( f_{sr} \) is greatly improved by removing the silicon underneath the inductor. The spiral inductors have a peak Q-factor of 35, and the maximum resonance frequency of the inductors is about 16 GHz. Fabrication of 2.5 turns inductor using MEMS technology is carried out. Measurements and characterization results are presented in this work.

Key-Words: Spiral inductor, MEMS technology, High Q-factor, RF applications.

1 Introduction

Recently, with the rapid growth of the demands in wireless communication products such as mobile phones and wireless network, low cost and high performance on-chip radio-frequency devices are strongly needed. One important limitation in achieving higher levels of integration and further reduction of fabrication costs in the front-end of microwave transceivers is set by the difficulty of achieving high-Q on-chip inductors [1]. Usually, the planar spiral inductors are integrated on the low-cost standard (low resistivity) silicon substrate using standard silicon technology and aluminum (Al) metal interconnects. These inductors exhibit poor Q-factor due to the severe substrate loss of the standard silicon substrate at microwave frequencies and the ohmic loss of the aluminum thin-film. As a result, novel low-cost technologies need to be introduced for fabricating silicon-based high-Q inductors for the high-performance single-chip RFICs. While the ohmic loss can be reduced by using high-conductivity metals such as Copper (Cu) [2] or gold [3], the reduction of the substrate loss remains the major obstacle for high-performance silicon-based inductors. To overcome the substrate loss, several approaches have been implemented with improved Q-factor. Ground-shielded inductors have been devised to reduce the substrate loss and noise coupling [4] with limited improvement in Q-factor (up to ten). Suspended inductors with the inductor metal separated from the lossy silicon substrate have offered Q factor as high as 50 at 7 GHz [5]. Another approach to reduce the substrate coupling is to insert a low-loss low-K dielectric layer between the inductor metal and the lossy silicon substrate [6]-[8]. Recently, the use of silicon micromachining techniques to remove the substrate underneath the planar inductors has significantly increased both the inductor self-resonant frequency \( f_{sr} \) and quality factor \( Q \) [9]-[12].

In this paper, a complete library of spiral inductors on silicon substrate is designed and modeled. Also fabrication of one of these designed inductors is done using MEMS technology. Post processing step is utilized to remove the silicon underneath the inductor using micromachining technique. Great improvements in the Q-factor and resonance frequency have been achieved by etching away the silicon underneath the spiral inductor. The HFSS 3D electromagnetic simulation results are presented. Measurements and characterizations of the fabricated inductor is performed and presented in the paper.

2 Inductor Design and Model

Various rectangular spiral inductors have been designed using Greenhouse’s method [9] and simulated using HFSS, a 3D electromagnetic simulator. The spiral inductors have a 2 \( \mu \)m aluminum line thickness and 150 \( \mu \)m inner diameter. Table 1 lists the number of turns, line spacing, and line width for these different spirals.
Table 1 Spirals different geometries.

<table>
<thead>
<tr>
<th>Device Number</th>
<th>N number of turns</th>
<th>S line spacing (µm)</th>
<th>W line width (µm)</th>
<th>Di Inner Diameter (µm)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>3.5</td>
<td>5</td>
<td>10</td>
<td>155</td>
</tr>
<tr>
<td>2</td>
<td>3.5</td>
<td>5</td>
<td>20</td>
<td>155</td>
</tr>
<tr>
<td>3</td>
<td>3.5</td>
<td>5</td>
<td>30</td>
<td>155</td>
</tr>
<tr>
<td>4</td>
<td>3.5</td>
<td>5</td>
<td>20</td>
<td>155</td>
</tr>
<tr>
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<td>3.5</td>
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<td>155</td>
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<td>6.5</td>
<td>5</td>
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<tr>
<td>11</td>
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<td>5</td>
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<td>155</td>
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<tr>
<td>12</td>
<td>3.5</td>
<td>5</td>
<td>20</td>
<td>155</td>
</tr>
<tr>
<td>13</td>
<td>3.5</td>
<td>5</td>
<td>20</td>
<td>180</td>
</tr>
</tbody>
</table>

The S-parameters which are calculated using HFSS are transformed into the Y-parameters from which the inductance \( L \) and quality factor \( Q \) can be calculated based on the following equations [13], respectively:

\[
L = \text{Im}(1/Y) / 2\pi f \tag{1}
\]

\[
Q = \text{Im}(1/Y) / \text{Re}(1/Y) \tag{2}
\]

Where: \( Y \) are the Y-parameters and \( f \) is the signal frequency.

### 3. Simulation Results

There exists a trade-off between the inductance and quality factor when increasing the number of turns of a spiral inductor. Figure 1a and b shows the variation of inductance \( L \) and quality factor \( Q \) due to the change in the number of turns for three spiral inductors of the same dimensions. It can be seen from the figure that, when the number of turns varies from 2.5 to 6.5, the inductance increases while \( Q \) and self resonance frequency \( SRF \) decrease. Increasing the number of turns will increase the inductor length \( l \). This results in increasing the self inductance. Moreover, the positive mutual coupling between the inductor turns will be increased as well. Both effects will cause the inductance to increase as the number of turns increases. The \( Q \) will be reduced due to increasing the series resistance, \( R_s \). The increase in \( R_s \) is due to increasing the inductor length. Second, the induced eddy currents in the inductor segments will increase as a result of increasing the number of turns. As the area of the inductor increases due to increasing the number of turns, the overlap capacitance \( C_s \) and the capacitance in the oxide layer \( C_{ox} \) are increased. As a result, the \( SRF \) will be reduced. Figure 2 shows the variation of inductance and \( Q \) for the simulated inductor groups as the line spacing increases. It can be seen from figure 2a and b that, increasing the line spacing from 5 µm to 25µm has a week effect on the total inductance and \( Q \). The \( SRF \) is kept constant as the line spacing varies. The main reason behind this is the small range which has been investigated for the line spacing. However, this range of line spacing is the practical range which is usually used in practical spiral inductors. The spiral size is usually limited by the chip area, which in turns prohibits the spiral line spacing to be greater than 25 µm.

![Fig. 1a Inductance variations with different No of turns](image1a)

![Fig. 1b Quality factor variations with different No of turns](image1b)
Figure 3 illustrates the effect of the line width on $Q$ for inductors with the same inductance but different line width. Three spiral inductors are designed with line width equal to 10, 20, and 30 µm. Inductors with wider lines have smaller series resistance $R_s$, which is inversely proportional to the width of the strip. However, they also have more shunt substrate parasitics because they occupy larger area. At low frequencies, the larger inductors offer higher $Q$ because of lower series resistance $R_s$. At high frequencies, the substrate effects as well as the proximity effects dominate and the smaller inductors actually achieve higher $Q$.

The variation in the wire cross-section dimensions has little effect on the inductance. Generally, wires with smaller cross-section have a slightly larger inductance because they generate more magnetic flux external to the wire.

Therefore, increasing the line width will slightly decrease the self inductance and also the mutual inductance for the spiral inductor. As the area of the inductor increases due to increasing the line width, the overlap capacitance $C_{ov}$ and the capacitance in the oxide layer $C_{ox}$ will increase. As a result, the SRF will be reduced.

Figure 4 illustrates $L$ and $Q$, SRF as a function of frequency when inner dimensions are varied, while all other characteristics are kept constant. The conclusion drawn is that as the area occupied by the spiral inductor increases, the inductance value increases, while the value of $Q$ and SRF decrease due to the reasons discussed previously.
4. Fabrication
Realization of 2.5 turn suspended inductor is achieved using standard technology MOSIS, 0.8um and CMP as a third party for etching process of the fabricated chip. Figure 4a shows the inductor’s layout. Etching simulation has been performed, to insure the required etching time before doing the post processing step. TMAH has been used to etch the silicon under the inductor to release the suspended structure. Figure 4b shows a scanning electron microscope (SEM) microphotograph for the fabricated suspended inductor after performing the etching step.

4 RF Measurement & Characterization
The two-port S-parameters of the fabricated inductor is measured using Network Analyzer and the SUSS-Micro-
5 Conclusion and Discussion
In this work a complete Library of spiral inductors with different line width, line spacing and number of turns have been designed and simulated. The performance of suspended spiral inductors on silicon substrate has been investigated using electromagnetic simulator. Fabrication of 2.5 spiral inductor is realized using MEMS Technology. The substrate material below the inductors is removed by wet etching process. The results show that the self resonant frequency $f_{sr}$ and $Q$ factor of the micromachined inductor increase with many factors as described in the designed inductors library. Also it is understood that significant inductor performance can be obtained with the proper selection of number of turns $N$, line width $W$ and line spacing $LS$.

References